THE R-9'ER

One-Tube Preamplifier Automatically Matches Antenna

**ELECTRICAL CIRCUIT**

![Circuit Diagram of the R-9'er](image)

Circuit Constants:
- $C_1$, $C_2$ = 5 mmfd fixed ceramic
- $C_3$, $C_4$ = 100 mmfd variable (Candwell EZ-100 AB)
- $C_6$, $C_7$, $C_9$, $C_8$ = 500 mmfd 400 volt mica
- $L_1$, $L_2$ = Slug-tuned ceramic form (Millen No. 60904)
- 10 meters—16 turns No. 26 enam core wound
- 6 meters—8 turns No. 26 enam core wound
- $R_1$, $R_2$ = 7000 ohm, 1/4 watt
- $R_3$ = 200 ohm, 1 watt
- $R_4$ = 15,000 ohm, 1/4 watt
- $R_5$ = 25,000 ohm, 4 watt potentiometer (Mallory M25MP)
- $R_8$ = 10,000 ohm, 1 watt
- $S$ = DPDT wafer switch (Mallory 3222)

**Fig. 1—Circuit Diagram of the R-9'er**

Are you having trouble picking those weak DX signals out of the noise? The R-9'er, using a single General Electric 6AK5 miniature tube, is designed to do exactly that. The R-9'er is an electronic impedance-matching device and a broad-band preamplifier, designed to work primarily on the 6 and 10 meter bands.

**Performance Characteristics**

The gain which can be achieved by this unit depends upon how well your antenna is matched to your receiver, but the minimum gain which may be expected is 30 decibels—about 5 R's! This gain comes about in two ways. The R-9'er, once it is tuned, automatically matches your receiving antenna to your receiver. In the usual ham shack, this problem is not given much consideration, but a tremendous gain can be obtained by a proper match. The problem is doubly important on the 6 and 10 meter bands, as at these frequencies the input impedance of the receiver may vary widely from its stated value. For example, a widely known communication receiver, stated to have an input impedance of 250 ohms, actually had an input impedance of 1500 ohms on 10 meters.

Tests made recently show that the average gain experienced, merely by properly matching the receiving antenna, is from several db to as high as 30 db.

In addition to this gain, the 6AK5 miniature tube acts as a broad-band r-f amplifier stage, giving an additional gain of approximately 30 db. This tremendous gain is possible only because of the electrical characteristics of the...
6AK5. This tube has a transconductance of 3000 microhams, which means that a voltage gain of approximately 35 can be achieved with a plate load of 7000 ohms, as used in the R-9'er. This amount of gain has been available only by former tubes at narrow band widths and with higher noise levels. The General Electric 6AK5 has been designed to give these high gains at wider band widths and at lower noise levels.

Here then is what the R-9'er will do for you—60 decibel gain (or more) if your present receiving antenna is not matched, or, assuming it is perfectly matched, a 30 decibel gain. In tests conducted at W2RDL's shack, the R-9'er brought in signals which could not ordinarily be heard even with the use of the BFO!

CIRCUIT DETAILS

Referring to Fig. 1, the circuit consists essentially of a broad-tuned grid and broad-tuned plate circuit, a standard cathode bias system, and an adjustable screen supply. The grid and plate circuits are identical except that capacitor C3 is employed as a plate blocking capacitor so that the plate tuning capacitor may be grounded.

In the grid circuit, capacitors C1 and C4 form the impedance matching network. A regular two-wire transmission line from the receiving antenna is brought to the input terminals, or a single wire antenna may be used and connected to the input lead which connects to the junction of C1 and C4. Inductance L1 must be tunable so that resonance may be achieved after C4 has been adjusted to match the antenna. Once C4 and L1, as well as C3 and L3, have been set, no further tuning is required for operation on that particular band.

With the constants shown, the R-9'er will match any input and output between 16 ohms and 2700 ohms. This may be calculated:

\[
\text{Impedance} = \frac{7000}{C_4 (C_1 + C_4)}
\]

The same formula may be applied to the plate side by substituting C, for C, and C' for C4.

All constants given must be strictly adhered to in duplicating the R-9'er, as even the values of the bypass capacitors are important. R, and R, must be 2000 ohms, as the band-width will be altered and the impedance formula changed if different values are used.

The band-width of the R-9'er with the constants as shown is approximately two megacycles on ten meters (28-38 mc) and five megacycles on six meters (58-55 mc), dropping off only one or two db at each end of the band when it is peaked in the center of that bandwidth.

The plate voltage is not critical, and any voltage available in your receiver will operate the 6AK5 satisfactorily.

CONSTRUCTIONAL DETAILS

The R-9'er is built in a 3 by 4 by 5 inch box, with all component parts mounted on the front panel. Figs. 3 and 4 show the essential details of construction. The switch, S, and the potentiometer, R2, are the two controls on the upper part of the front panel, with capacitors C and C3 being mounted directly beneath.

The coil box occupies the central portion of the box, and is so arranged that the main support on the coil form, a piece of \(\frac{3}{4}\) inch by \(\frac{3}{8}\) inch aluminum, \(\frac{3}{4}\) inch thick, just fits into the central shield on the box, which is made of \(\frac{1}{4}\) inch thick aluminum. With the coil plugged into the R-9'er, a solid shield is thus formed which completely isolates the grid section from the rest of the circuit. It is very important to have complete shielding between grid and plate. The polystyrene base on the coil is \(\frac{1}{4}\) inch by \(\frac{3}{8}\) inch, and the aluminum front of the coil measures 2 inches by \(\frac{1}{4}\) inch. One corner is cut on the polystyrene base in order to provide a method of keying the coils for proper insertion. The cutout in the panel is similarly keyed. The coil forms are mounted on a thin piece of aluminum (see Fig. 5 of page 4) so that the center of the grounding strip contacts a grounding spring mounted on the \(\frac{1}{4}\) inch aluminum shield. This grounding spring is identical to the one shown in Fig. 3 which is mounted on the rear of the shield. The purpose of the latter spring is to contact the inside of the box, in the rear, for good grounding.

The pips on the coil are Millen No. 10029, which fit into two crystal sockets (Millen No. 38002). These sockets are mounted on the \(\frac{1}{4}\) inch wide aluminum shield, as may be seen in Fig. 4.

(Continued on last page)
QUESTIONS AND ANSWERS

Do you have any questions about tubes or tube circuits? Lighthouse Larry would like to answer them for you. For each question published you will receive $10 worth of Q.E. electronic tubes. All questions not published will be answered prompt-ly by mail. Send your questions in Lighthouse Larry, Tube Division, Bldg. 269, General Electric Company, Schenectady, New York, or in Canada, to Canadian General Electric Company, Ltd., Toronto, Ont.

Question: Considering the case of a tube used as a Class C r-f amplifier, does the amplification factor (mu) of the tube bear any relationship to the driving power required?—W9KFN.

Answer: There is no simple relationship between the mu of a tube and the driving power required. Practical considerations in tube construction result in the fact that the optimum mu for triodes in Class C operation is about 20. In other words, a higher or lower mu tube might require more driving power than a tube with a mu of 20. A comparison of tube transconductance (Gm) is a better guide to driving power, as a tube with a high transconductance will require less drive than a tube with a low transconductance, if tubes with similar amplification factors are considered.—Lighthouse Larry.

Question: When using a neutralized triode with combination grid resistor and battery bias, is there an optimum ratio of battery voltage to resistance-developed voltage?—W1OJC.

Answer: It is important to use sufficient fixed (battery of power-supply) bias to limit the plate current in the event that excitation fails. To use more fixed bias than this is not economical. In addition, use of grid resistor bias has several advantages. The resistor acts as a current limiting device if a grid-to-cathode arc-over takes place. The grid resistor also prevents the grid of the tube from being driven to a high positive voltage. The best rule to follow is to provide enough fixed bias to keep the plate dissipation within ratings in case of an excitation failure, and obtain the remainder of the required bias from a resistor.—Lighthouse Larry.

TRICKS AND TOPICS

How did you solve that last problem that almost had you stumped? Is it about tubes, antennas, circuits, etc., Lighthouse Larry would like to tell the rest of the hams about it. Send in lid For each "trick" accepted you win $10 worth of Q.E. Electronic Tubes. No entries returned. Submit to Lighthouse Larry, Tube Division, Bldg. 269, General Electric Company, Schenectady, New York or in Canada, to Canadian General Electric Company, Ltd., Toronto, Ont.

METAL CHASSIS PRESERVER

After building up a few very nice looking units on cadmium-plated chassis pans a few years ago, I was very disappointed to find that they had turned dark in color and had rusted in spots.

When the bands re-opened, I decided to rebuild, using new pans. I drilled and punched all the needed holes in the new cadmium pans, and used a ball of absorbent cotton soaked up in carbon tetrachloride to wipe the finished job all over, inside and out.

Then with a nice clean piece of cotton rag, I waxed the chassis inside and out, using regular paste auto wax. I then mounted the parts, even going so far as to soak up the little paper gaskets between the porcelain feed-through insulators with the wax.

After the parts were all mounted and the unit tested, I gave it another light going-over with the wax rag, without wiping off for the polishing effect.

My rig is in the basement, and these treated chassis pans look better now than the day I got them from the local distributor, in spite of the dampness that is attributed to basements.—W9GPI.

(Ed. Note: This trick really does the job, but care should be taken that the chassis is not handled after the final polishing, as the fingerprints may rub off enough wax to expose the chassis so that it will eventually rust.—L.L.)

RESONANT RECEIVER CHOKE

I ran across the following trick recently that helped me out of a bad situation. In the design of a high gain UHF superheterodyne receiver I was found necessary to use filament chokes to prevent regeneration. Inasmuch as the intermediate frequency was 12 megacycles, difficulties arose in trying to wind chokes to do the job at that frequency, because of the low distributed capacity of the choke and the high inductance required.

This problem was solved by making a resonant choke with a fixed capacitor in the following manner. A short piece of bakelite tubing was procured and a ceramic capacitor inserted in the tubing. The choke was then wound over the tubing, and connections made to the capacitor at each end of the coil, thus connecting the coil and capacitor in parallel.

For 12 megacycles the construction is: Tubing one inch long, ¼ inch O.D. and ½ inch I.D. Capacitor, 20 mmf. Coil No. 30 enameled wire close-wound for ¾ inch.

In use, one choke is connected to each tube, right at the socket. The other filament lead is common.—Morris Allen.
The G-E 6AK5 tube is mounted horizontally. Fig. 3 and 4 show how the grid pin on the tube socket projects on one side of the shield with the remainder of the pins on the other side of the shield. Switch S is mounted on this shield. The input connections are mounted on a third shield which cuts through the center of switch S, shielding the input and output circuits.

Placement of parts is not too critical if adequate shielding is maintained. Lack of shielding may cause unwanted regeneration and possible spurious oscillations.

**OPERATING ADJUSTMENTS**

Input and output connections should be made to the R-9'er with well-insulated wire, preferably coaxial cable. Switch S should be set so that the amplifier is cut out, and the receiver tuned to a signal in the approximate center of the band. A local signal is preferable. The amplifier should then be cut in by the switch, the screen potentiometer adjusted to give maximum voltage, and the grid condenser C1 tuned together with L1 until the signal is heard. The signal should then be peaked with an R-meter or an output meter by tuning L1, adjusting C1, returning L1, readjusting C1, etc., till the signal is maximum. This process should be repeated with the plate circuit, C2 and L2.

If C1 is found to be at full maximum or minimum capacity, the length of the antenna feeder must be altered. Conversely, the length of the line between the R-9'er and the receiver must be altered if C1 does not tune near its middle capacitance. To correct this situation, add a quarter-wave length of line and prune this line until the capacitor peaks the signal at approximately center scale. For 6-meter operation the output line should be as short as possible, to ensure minimum capacitance on the output side.

After the entire unit has been peaked, the screen potentiometer (R1) should be adjusted for maximum output, keeping the voltage on the 6AK5 screen as low as possible, with output as high as possible. Once all adjustments are made for both coils, it is only necessary to peak capacitors C2 and C1 when changing bands, as the coils remain at resonance after once being adjusted.

Coil data for L1 and L2 is given for only 6 and 10 meter operation, although the unit will operate on any band. However, most antennas match into the receiver reasonably well at 14 megacycles and lower frequencies, so that the R-9'er does not appear to give as much gain at the lower frequencies.

![Fig. 5—View of R-9'er Cabinet (Note that Cabinet is Mounted on Polyethylene Base)](image)

**6AK5**

**BASING DIAGRAMS—PIN CONNECTIONS**

![Diagram](image)

**PARASITICS** On page four of Ham News, Vol. 1, No. 3, the tube basing diagram shown for the 5855 should be used for the 12BA6 and vice versa.

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_Lighthouse Larry_

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